



OPEN Learning uplinks and downlinks transmissions in RF-charging IoT networks

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This paper considers uplink and downlink transmissions in a network with radio frequency-powered Internet of Things sensing devices. Unlike prior works, for uplinks, these devices use framed slotted Aloha for channel access. Another key distinction is that it considers uplinks and downlinks scheduling over multiple time slots using only causal information. As a result, the energy level of devices is coupled across time slots, where downlink transmissions in a time slot affect their energy and data transfers in future time slots. To this end, this paper proposes the first learning approach that allows a hybrid access point to optimize its power allocation for downlinks and frame size used for uplinks. Similarly, devices learn to optimize (1) their transmission probability and data slot in each uplink frame, and (2) power split ratio, which determines their harvested energy and data rate. The results show our learning approach achieved an average sum rate that is higher than non-learning approaches that employed Aloha, time division multiple access, and round-robin to schedule downlinks or/and uplinks.

Future Internet of Things (IoT) networks will consist of low-power devices that sense their environment and transmit data to a gateway¹. The gateway may then use the data from devices to train a neural network². Further, a gateway may instruct devices to carry out sensing task(s)³ or control an actuator. In these scenarios, channel access is a key issue in order to facilitate uplinks and downlinks transmissions over the same channel. Moreover, devices may experience collision when they upload their data to a gateway. Hence, a key issue is to determine when a device accesses a channel given unknown number of contending devices.

Another key issue is managing the available energy of devices, where they rely on a hybrid access point (HAP) for energy. Briefly, these devices are charged via far-field wireless charging; see⁴ for an example prototype. Specifically, radio frequency (RF)-charging takes advantage of the existing spectrum that is used for data transmissions to also deliver energy. This fact has led to technologies such as Simultaneous Wireless Information and Power Transfer (SWIPT)⁵, where devices are able to receive both information and energy simultaneously. In this respect, SWIPT supports time switching⁶ and power splitting⁷. Specifically, with a power splitter, a receiver divides the power of a received signal between its energy harvester and data decoder. This division of power is a variable to be optimized by the receiver. In contrast, time switching has two phases. In the first phase, the HAP charges devices. After that, in the following phase, devices are allocated a given time slot to transmit data. The main variable to be optimized is the time allocated to each device for uplink transmission as well as the charging phase duration used by the HAP⁸.

To address the aforementioned issues, this paper considers joint optimization of uplink and downlink data communications in a RF-energy harvesting IoT network. Further, it takes advantage of non-orthogonal multiple access (NOMA)⁹ for downlink transmissions, where a HAP uses superposition coding to transfer information to all devices. For uplinks, past works pre-assign time slots or sub-carriers for each user, which will cause slot wastage when users run out of battery. Considering that users transmit opportunistically based on their energy level, this paper adopts random channel access, framed slotted Aloha, to avoid wasting pre-assigned time slots. Moreover, the HAP uses Successive Interference Cancellation (SIC) to decode concurrent transmissions from devices. Note that NOMA allows for higher spectrum efficiency, and thus it is a key technology in future networks¹⁰. In terms of energy delivery, devices adopt power splitting. Note that there is a trade-off between the harvested energy and information decoding, which affects uplink and downlink rates, respectively. Concretely, less harvested energy results in less uplink transmission power which may impair the uplink rate. Accordingly, less power directed to information decoding impairs the downlink rate. Thus, a key problem is to determine a suitable power split ratio that maximizes the amount of harvested energy and sum rate during downlink transmissions. Lastly, the HAP employs Frame Slotted Aloha (FSA) for uplink transmissions, meaning devices are not allocated a fixed time slot. A key advantage of FSA is that an HAP does not have to allocate a fixed

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time slot to devices with insufficient energy to transmit. In this respect, the HAP can optimize its frame size in accordance with the number of transmitting devices.

Figure 1 shows the downlink and energy delivery process in an example IoT network with RF-charging devices. During downlink, the HAP simultaneously transmits data and RF energy to all devices, where the HAP superposes its transmission to both devices. For ease of exposition, assume that the HAP uses 2 W and 1 W when it transmits to U_1 and U_2 , respectively. Further, assume the channel power gain between the HAP and devices equals one. Lastly, each device has a power split ratio of θ . As shown in Fig. 1a, device U_2 adopts a power split ratio of $\theta = 0.8$. Thus, a total of $0.8 \times 3 = 2.4$ W is sent to its information decoder. The remaining 20% of its received power is sent to its energy harvester. Device U_1 adopts a power split ratio of $\theta = 0.5$. This means device U_1 sends 1.5 W to its information decoder and the other 1.5 W to its energy harvester. By using SIC decoding, device U_1 and U_2 are able to iteratively decode the signal.

Figure 1b demonstrates downlink and uplink transmissions for two time frames. The HAP first superposes all data together to all devices during the downlink period. After that, devices use their harvested energy to transmit during the uplink period. To do so, in frame $t = 1$, device U_2 selects the first data slot while device U_1 selects the third data slot. In this case, both data transmissions are successful. However, in frame $t = 2$, devices selected the same data slot. In this case, due to SIC, their transmission is also successful.

There are a number of challenges. First, the uplink transmit power of devices is a function of their harvested energy in prior time slots or downlinks from the HAP¹¹. Second, the energy level of devices depends on past channel gains, power split ratio and transmissions. Third, information is causal, meaning both HAP and devices know the current and past channel gains information only. Consequently, devices are unable to predict their future channel states or future energy arrivals, which undoubtedly increases the difficulty for them in making transmission decisions. Specifically, devices do not know whether they should reserve their precious energy for future slots with better channel gains or should transmit immediately. Fourth, the HAP is unaware of the number of contending devices and the energy level of devices. In practice, obtaining this information involves signaling, which consumes the precious harvested energy of devices. To address these challenges, this paper utilize Q-learning based approach to learn the system energy arrivals and channel condition variation. Henceforth, this paper makes the following contributions:

- It studies an IoT network that uses NOMA and FSA. It addresses a novel problem that aims to jointly maximize uplink and downlink sum rates over *multiple* time frames using only causal information. To the best of our knowledge, no prior works have considered a system that employs FSA for uplinks nor the same problem. Further, they have not addressed the said challenges jointly; see “[Related works](#)” Section for details.
- It shows how the uplink and downlink transmission problem can be modeled as Markov Decision Process (MDP). Advantageously, the MDP is model-free, meaning it does not require statistics of an environment beforehand. This means it only needs to observe the current system state, and then executes an action as per a learned policy. In this respect, to determine the optimal policy, this paper outlines Multi-Q; it is the *first* reinforcement learning approach for the problem at hand and system. It yields a communication policy for different channel conditions, where for each system state, it determines the HAP’s transmit power allocation and frame size. Further, it trains devices to use the correct power split ratio that maximizes their harvested energy and downlink sum rate. Lastly, devices use Multi-Q to determine a slot in a given frame and transmission probability.
- It presents the first study of Multi-Q. The simulation results show that Multi-Q achieves an average sum rate of 44 b/s/Hz which is 6× that of Aloha, 2.3× that of time division multiple access (TDMA), and 30% more than round-robin.

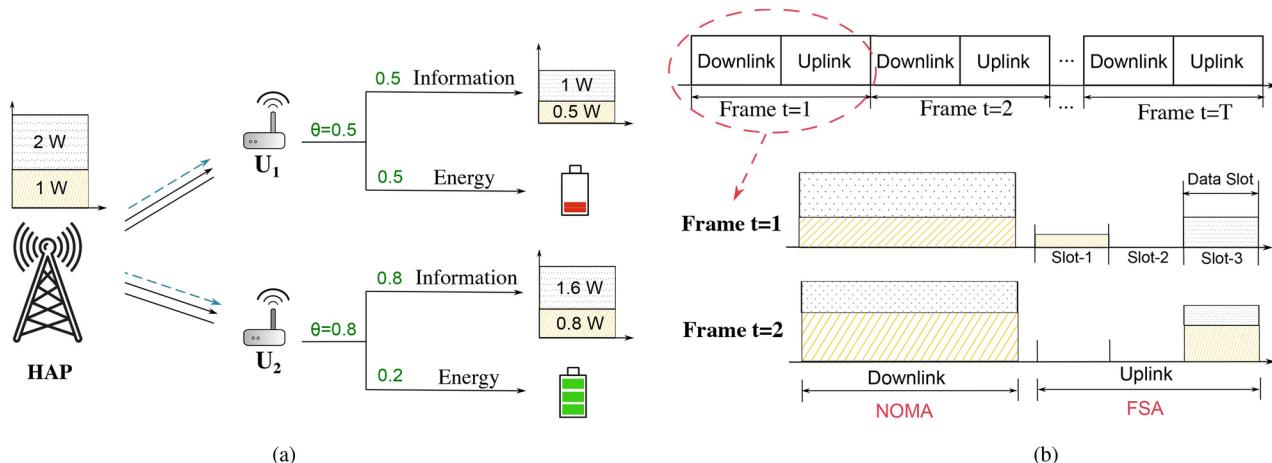


Fig. 1. An example of downlink and uplink data transmissions.

Related works

In general, past works that consider downlink and uplink communications in RF-charging networks optimize over *one* frame; i.e., they do not consider energy evolution and future channel gains of devices. These works mainly consider two different kinds of frame structures. One frame structure, see e.g.,^{12–14}, is where the HAP first transfers energy to all users. After that, each user is assigned a distinct time slot during downlink. The HAP then sequentially sends data to each user. Works such as¹³ consider users that not only harvest energy during downlinks via a time switching strategy, but they also harvest energy whenever there are uplink transmissions. The work in^{14–17} considers users who harvest energy when the HAP transmits information to other users. During their assigned slot, a user employs power splitting to harvest energy. Lastly, in reference¹⁵, the authors also consider an interfering source. Users harvest energy from both their HAP and interfering signals via power splitting.

Some works consider an HAP that simultaneously sends data to all users using NOMA or multiple-input multiple-output (MIMO) technologies. Example works include^{18,19}, where the HAP uses a MIMO system. Qin et al.¹⁸ assume a fixed power split ratio and time division duplex. In a subsequent work, i.e.,¹⁹, they optimize the power split ratio for each user to maximize system throughput. References^{20–22} consider NOMA for downlink. Li et al.²⁰ employ NOMA in both single-input single-output (SISO) and MIMO models. In the case of SISO-NOMA, the authors optimize the power split ratio for each user. In the case of MIMO-NOMA, except for determining the power split ratio at devices, the authors also optimize the time allocation for downlink and uplink duration to enhance sum rate. Baidas et al.²¹ jointly optimize time switching and power allocation in a single-cell NOMA system to maximize the sum rate of uplink and downlink while ensuring quality of service requirements of users. In a subsequent work²², Baidas et al. further consider a NOMA system with clusters of users. The aim is still to maximize the sum rate of uplink and downlink while ensuring quality of service requirements are met. The authors jointly optimize time switching and power allocation of each cluster, and its sub-carrier assignment.

Another research direction is to adopt different sub-carriers for uplink and downlink communications. For example, Rezvani et al.²³ consider a multi-user orthogonal frequency division multiple access (OFDMA)-based system with one base station and one local access point, where the base station can offload data to a local access point. The aim is to maximize uplink throughput subject to a minimum required downlink data rate of each user. To do this, the authors optimize power split ratio at users, joint sub-carrier allocation, and transmit power allocation. Na et al.²⁴ categorize sub-carriers into two groups for information decoding and energy harvesting, respectively. Xiong et al.²⁵ aim to jointly optimize the downlink and uplink energy efficiency and prolong system lifetime in a Time Division Duplex (TDD) Orthogonal Frequency Division Multiple Access (OFDMA) system with a power split strategy.

Table 1 highlights the novelties of our work. Briefly, many works have employed convex optimization and solves a deterministic problem, i.e., they do not consider imperfect or stochastic channel gains. For example, the work in^{12,13,17–20} casts or transforms their proposed maximization problem into a concave one and use convex program. Some other works also consider Mixed-Integer Non-Linear Program (MINLP), e.g., reference^{21–23}. To date, only the work in²⁶ has considered multiple time slots with imperfect channel gains. Lastly, in²⁶,

References	Working principle	SWIPT	Downlink	Uplink	Non-linear energy model	Energy evolution	Learning	Decision variables
^{12–14}	HAP first charges users followed by data transmission to each user.	TS	TDD	TDD	✗	✗	✗	Time allocation Transmission power
^{15–17}	Users harvest energy whenever HAP transmits data to users.	PS	NOMA TDMA	NOMA-TS TDMA	✗	✗	✗	Time allocation Transmission power Beamformer
^{18,19}	HAP transmit energy and data in multiuser MIMO system.	PS	TDD	TDD	✗	✗	✗	Beamformer Time allocation Power split ratio
^{20–22}	HAP transmits energy and data to users using NOMA.	PS TS	NOMA	NOMA	✗	✗	✗	Time allocation Power splitting ratio
²³	An OFDMA system with uplinks and downlinks sub-carriers.	PS	FDD/TDD	FDD/TDD	✗	✗	✗	Subcarrier allocation Transmission Power Power splitting ratio
²⁴	Different sub-carriers for energy delivery and data transmissions.	N/A	OFDM	NOMA	✗	✗	✗	Sub-carrier allocation Transmission Power Power splitting ratio
²⁵	Different sub-carriers for energy harvesting and data transmissions.	N/A	TDD	TDD	✗	✗	✗	Transmit Power
²⁶	High and low frequency bands for energy and data transmissions.	N/A	TDD	TDD	✓	✗	✗	Transmit power Charging duration
This work	HAP transmits data and energy to users using NOMA, and receives data from devices over the same channel.	PS	NOMA	Aloha-SIC	✓	✓	✓	Transmit power, Power split ratio, Transmit probability, Slot selection, Frame size

Table 1. Comparison between joint uplink and downlink communication networks. TS and PS denote time switching and power splitting, respectively.

transmissions and receptions are carried out using TDD. A key innovation is the use of high and low frequency bands for energy and data transmissions. Further, access point optimizes its beamforming weight according to channel condition and energy at devices.

Our work fills a number of gaps. First, unlike past works that assume non-causal information and perfect channel information, we consider the causal case, meaning the HAP and devices make decision without requiring future channel gains information. Second, these works do not consider random channel access. Specifically, they consider pre-assigned time slots or sub-carriers, see^{12-14,23-25}. Thus, there will be wasted slots due to device energy outage which will impair transmission efficiency and system throughput. In contrast, for practical reasons, our devices employ slotted Aloha to transmit to a SIC-capable HAP. Slotted Aloha is a random channel access method that enables flexible transmissions based on battery states. With SIC enabled, collisions can be resolved to further improve system throughput. Third, they optimize resources over one slot. Specifically, except for Yao et al.²⁶, they do not consider energy evolution and future channel gains of devices nor the coupling between energy level across time slots. We note that Yao et al.²⁶ consider a known probability distribution of channel and data arrivals. Further, they do not consider random channel access and do not aim to maximize system throughput.

System model

A HAP serves N energy harvesting devices; each device is denoted as U_i , where $i \in \{1, 2, \dots, N\}$. The HAP uses NOMA in the downlink and devices employ FSA for uplinks, where devices select one of M slots to transmit their packet. Time is divided into frames and indexed by t . At the beginning of each frame, the HAP will send pilot symbols for channel estimation. After that, each frame is divided into a *downlink* and *uplink* period, which respectively has length τ_d and τ_u . During the downlink period, devices employ power splitting⁵ to split received power into two parts, namely energy harvesting and information decoding. After that, there are M time slots for uplinks.

Channel model

We consider Rayleigh block fading channels²⁷. The channel remains the same within one frame but varies across frames. Let d_i be the distance between U_i and the HAP, n denotes the path loss exponent, λ denotes an exponential random variable with unity mean, and h_i^t denotes the channel gain between user U_i and the HAP in time frame t . The channel power gain h_i is defined as $h_i^t = \lambda d_i^{-n}$ ²⁸.

From a practical point of view, we consider causal channel information. That means HAP and devices make transmission decisions only with the current and past channel gains information. Consequently, even Rayleigh fading drives the channel state variation, neither HAP nor devices are aware that state transitions are driven by a Rayleigh distribution. Concretely, for a given time slot, devices cannot predict any future channel states or energy arrivals. Hence it is hard for a device to decide whether it should use up its energy to transmit or it should reserve its energy for future slots with a better channel state.

Downlink

During each downlink period, the HAP superposed all signals together and transmit the resulting composite signal to all users¹⁰. The HAP has a maximum transmit power of P , and the power allocated for user U_i at time frame t is p_i^t , where $0 \leq p_i^t \leq P$. Moreover, the sum of transmit power to each user must not exceed P ; formally, $\sum_{i=1}^N p_i^t \leq P$. Further, each user U_i divides its received signal into two signals with a split ratio of θ , where $0 \leq \theta \leq 1$. Let θ denote the fraction of received power devoted to information decoding. The remaining $1 - \theta$ fraction of the received power is sent to an energy harvester.

Downlink information decoding

Users have a SIC decoder²⁹. Briefly, each user U_i starts its SIC decoding from the strongest signal by treating other signals as interference. After having successfully decoded the strongest signal, user U_i will subtract the decoded signal from the composite signal and proceeds to decode the next strongest signal. This continues until user U_i decodes its signal.

An example is shown in Fig. 2, where the HAP transmits with more power to user U_2 than U_1 . User U_1 decodes the signal designated for U_2 first and subtracts it from its received composite signal. After removing the signal from U_2 , user U_1 decodes its signal. As for user U_2 , it directly decodes its signal by treating the signal of user U_1 as interference.

Let n_0 denote the noise power and W denote the bandwidth. The achievable downlink rate for user U_i at time frame t is

$$\check{R}_i^t = W \log_2 \left[1 + \frac{\theta h_i^t p_i^t}{\theta h_i^t \sum_j^N p_j^t + n_0} \right]. \quad (1)$$

Energy harvesting

Each user is equipped with an RF-energy harvester, e.g., P2110B RF-energy harvester³⁰. Let \check{P}_i^t denote the received power at user U_i . It is calculated as $\check{P}_i^t = h_i^t P$. It transfers $(1 - \theta)P_i^t$ amount of power to its energy harvester. Note that the RF-energy conversion process is non-linear, which is a function of the received power. We consider a practical non-linear energy harvesting model³¹. Denote the energy conversion efficiency as η , which has range $[0, 1]$. It is calculated as

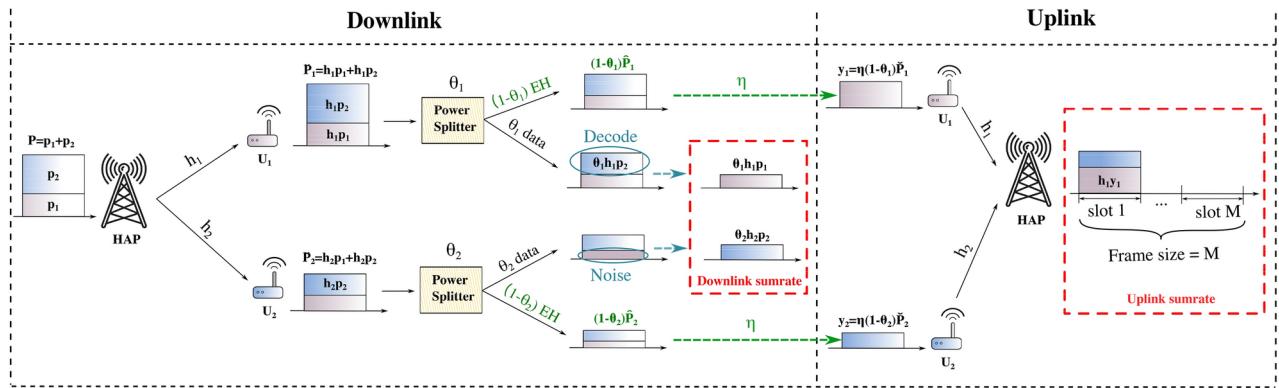


Fig. 2. An example of downlink and uplink transmission. There are two users and user U_2 has poorer channel condition and a higher transmit power. The left side of the figure shows downlink transmission, where the HAP transmits energy and data to users. Users use power splitting to harvest energy. The right side of the figure shows users transmit data via FSA in the uplink. The HAP uses SIC to decode information.

$$\eta = \frac{\chi_i^t}{\tilde{P}_i^t}, \quad (2)$$

where $\chi_i^t = (\Psi_i^t - M\Omega)/(1 - \Omega)$, $\Omega = 1/(1 + e^{ab})$, $\Psi_i^t = M/(1 + e^{-a(\tilde{P}_i^t - b)})$. Here, M is the maximum harvested power, and the value of a and b is as per a harvester's circuit.

Denote ξ_i^t as the harvested energy of U_i in time frame t . Formally, the harvested energy is

$$\xi_i^t = \eta(1 - \theta)\tilde{P}_i^t\tau_d. \quad (3)$$

Let v_i^t denote the amount of energy consumed by U_i for uplink transmission in frame t . Thus, each device has energy level E_i^t that evolves as per $E_i^t = E_i^{t-1} + \xi_i^t - v_i^t$. Moreover, each user has a battery capacity of B_{max} . This means if a device's battery is full, any subsequent energy arrival is lost. Consequently, the energy level of user U_i evolves as per

$$E_i^t = \min(B_{max}, E_i^{t-1} + \xi_i^t - v_i^t). \quad (4)$$

Uplink

Users use FSA for uplink transmissions. This means the HAP *does not* allocate a fixed time slot to a device. Concretely, users will randomly select a time slot to access the channel when it has sufficient energy. In contrast, if the time slot is pre-assigned to each device by the HAP, time slots may be wasted if any device experiences an energy outage. Moreover, if the HAP pre-assigns time slots based on the energy level of each device, it requires the HAP to gather energy level information. Consequently, it requires the HAP to poll devices, which is not practical when there are many devices. Further, this also wastes the precious energy of devices. In contrast, FSA provides energy harvesting nodes with more flexibility to report data. This means devices can transmit more flexibly based on their energy level to avoid wasting slots. Transmission efficiency will be improved since fewer time slots are wasted due to battery outages. For this reason, we adopt FSA, and have the HAP adjusts the frame size used for uplink transmissions based on system states.

The HAP has SIC capability²⁹, meaning it is able to decode multiple transmissions within a time slot. Let ε_0 be the minimum energy used to transmit one packet. Thus, user U_i will only transmit when its battery level E_i^t exceeds ε_0 . Further, user U_i will use up all its available energy to transmit. We denote the uplink transmission power of U_i as y_i^t . Then we calculate y_i^t as per $y_i^t = \frac{E_i^t}{\tau_{up}/M}$, where M is the frame size. For each user U_i , we record the number of transmissions in its selected slot as Φ_i^t . Therefore, the condition $\Phi_i^t \leq 1$ represents the fact that there are no other users transmitting in the same slot with user U_i . Otherwise, the condition $\Phi_i^t > 1$ means there are users transmitting in the same slot with user U_i . Consequently, the achievable uplink transmission rate for user U_i is defined as

$$\hat{R}_i^t = \begin{cases} W \log_2 \left(1 + \frac{h_i^t y_i^t}{\sum_j h_j^t y_j^t + n_0} \right), & \Phi_i^t > 1, \\ W \log_2 \left(1 + \frac{h_i^t y_i^t}{n_0} \right), & \Phi_i^t = 1, \\ 0, & \Phi_i^t < 1. \end{cases} \quad (5)$$

Problem

Given the aforementioned system, the goal is to optimize both uplink and downlink sum rate, i.e., summation of downlink rate R_i^t and uplink rate \check{R}_i^t . Here, the uplink rate R_i^t and downlink rate \check{R}_i^t are calculated as per Eqs. (1) and (5), respectively. Moreover, in each time frame t , a policy π returns all the parameters used in Eqs. (1) and (5). Specifically, a policy π returns the downlink transmission power p_i^t , uplink transmission probability ρ_i^t , uplink slot selection δ_i^t , frame size M , and power split ratio θ . Formally, a policy π is defined as $\pi = [p_i^t, \rho_i^t, \delta_i^t, M, \theta]$. Thus, the joint sum rate is calculated as per Eq. (6):

$$R_i^t(\pi) = \hat{R}_i^t(\pi) + \check{R}_i^t(\pi). \quad (6)$$

Define $\Pi = [\pi_1, \pi_2, \dots]$ as a collection of available policies. Our problem is to find the optimal policy $\pi^* \in \Pi$ that maximizes the following long-term cumulative joint uplink and downlink reward:

$$R(\pi^*) = \arg \max_{\pi \in \Pi} \frac{1}{T} \mathbb{E} \left[\sum_{t=1}^{\infty} \sum_{i=1}^N R_i^t(\pi^*) \right]. \quad (7)$$

To solve the optimal policy π^* , we need to determine the following quantities: (1) Downlink transmission power p_i^t of the HAP for each device U_i in frame t , (2) Uplink transmission probability ρ_i^t of device U_i , (3) Uplink slot selected by device U_i , namely δ_i^t in each frame, (4) Frame size M , and (5) Power split ratio θ of all devices.

MDP model and Q-leaning approach

We first show how the uplink and downlink process can be modeled as an MDP³². After that, we introduce conventional Q-learning³³. Note that Q-learning is a sequential decision approach that learns the optimal policy without using non-causal information. Advantageously, it is model-free, meaning they are able to learn the optimal policy by only observing system states over time. Specifically, Q-learning allows the system to learn the fact that channel state transitions are driven by Rayleigh distribution with only causal channel information. Then we introduce stateless Q-learning³⁴. Finally, we outline Multi-Q and show how it allows a HAP and users to use conventional Q-learning to learn the optimal policy that determines downlink power allocation, uplink transmission probability and slot selection in each frame. Moreover, Multi-Q also employs stateless Q-learning to determine the frame size for uplink transmissions and power split ratio of devices.

MDP model

To model the sequential decision process taken by the HAP and devices, we use an MDP model. It is defined as a tuple $[\mathcal{S}, \mathcal{A}, \mathcal{T}, \mathcal{R}]$. Here, the state space is denoted as \mathcal{S} . The action space \mathcal{A} includes a set of actions a . A policy π returns the action a for state s . After an agent takes action a in state s^t , the system will transition to a new state s^{t+1} with a transition probability of $\mathcal{T}(s^{t+1}|s^t, \pi(s^t))$. In addition, the agent obtains a reward $\mathcal{R}(s^{t+1}|s^t, \pi(s^t))$.

Our downlink MDP is defined as follows:

- *State*: A downlink state $\check{s} \in \check{\mathcal{S}}$ includes the channel conditions of all devices. Each state is defined as $\check{s} = [h_1, h_2, \dots, h_N]$.
- *Action*: The downlink action space is defined as $\check{\mathcal{A}} = [\check{a}_1, \check{a}_2, \dots]$. Each downlink action $\check{a} = [p_1, p_2, \dots, p_N]$ represents the downlink NOMA power allocation for all users at the HAP.
- *Transition probability*: We consider a model-free MDP model. Hence, the transition probability is unknown.
- *Reward*: The reward function $\check{\mathcal{R}}$ is the throughput of downlinks, see Eq. (1).

The uplink MDP is defined as follows:

- *State*: An uplink state $\hat{s}_i \in \hat{\mathcal{S}}$ includes the channel condition and battery level of user U_i . Formally, a state of user U_i is defined as $\hat{s}_i = [h_i, E_i]$.
- *Action*: An uplink action $\hat{a} \in \hat{\mathcal{A}}$ includes time slot selection δ_i and transmission probability ρ_i . Thus, an uplink action is defined as $\hat{a}_i = [\delta_i, \rho_i]$, which represents the fact that user U_i selects the slot indexed by δ_i^t and transmits with probability ρ_i .
- *Transition probability*: The transition probability between states is unknown.
- *Reward*: The reward function $\hat{\mathcal{R}}$ is the transmission rate of uplinks, see Eq. (5).

Q-learning

We employ two types of Q-learning methods, conventional Q-learning³³ and stateless Q-learning³⁴. Both conventional Q-learning and stateless Q-learning learn Q-values. However, conventional Q-learning learns Q-value for action and state pairs, while stateless Q-learning learns Q-value just for actions without any states.

Conventional Q-learning

Q-learning learns the optimal policy based on a Q-table. A Q-table is indexed by a state-action pair (s_t, a_t) , and returns the corresponding Q-value $Q(s_t, a_t)$. Each Q-value $Q(s_t, a_t)$ represents the expected discounted reward for taking action a_t in state s_t ³³. The aim of Q-learning is to calculate $Q(s_t, a_t)$ for each action and state pair. To learn the optimal policy, an agent first obtains its current state s_t . Secondly, it looks up its Q-table to find the corresponding Q-values for state s_t and selects the action a_t with the highest Q-value. After the agent selects

action a_t , the system will return a corresponding reward $r(s_t, a_t)$. Then the agent observes its next state s_{t+1} and finds the highest Q-value. Lastly, the agent updates its Q-table based on its obtained reward and the highest Q-value for the next state. We denote α as the learning rate factor, γ as the discount factor, where $\alpha, \gamma \in [0, 1]$. Concretely, Q-learning uses Bellman's equation to update its Q-table as per

$$Q(s_t, a_t) = (1 - \alpha)Q(s_t, a_t) + \alpha(r(s_t, a_t) + \gamma \max \check{Q}(s_{t+1}, a_{t+1})). \quad (8)$$

Stateless Q-learning

Stateless Q-learning³⁴ learns the optimal policy without any states. The stateless Q-table only contains the value of actions. We denote $\lambda \in [0, 1]$ as the stateless learning rate and the reward is denoted as $r(a)$. Thus, stateless Q-learning updates $Q(a)$ using

$$Q(a) \leftarrow Q(a) + \lambda(r(a) - Q(a)). \quad (9)$$

Under this stateless setting, an agent maintains a Probability Mass Function (PMF), denoted as

$$\Pr(a_i) = \frac{e^{Q(a_i)/T}}{\sum_a e^{Q(a)/T}}, \quad (10)$$

which calculates the probability of taking action a_i .

Multi-Q learning

Now we are ready to outline our proposed Q-learning approach, named Multi-Q, to solve Problem (7). Multi-Q is composed of three layers, namely the *uplink*, *downlink*, and *stateless*. Figure 3 shows the Multi-Q framework. The *downlink* and *uplink* layer adopt conventional Q-learning while the *stateless* layer employs stateless Q-learning. All layers use ϵ -greedy for action selection. Thus, initially, each agent has ϵ probability to randomly select an action. After that, we decay the value of ϵ to ensure convergence. Concretely, at the *downlink* layer, the HAP is the agent to learn the downlink MDP action $\check{a} = [p_1, p_2, \dots, p_N]$ which includes the power allocation for each user. The HAP starts with randomly selected power allocation first. During this warm-up period, the Q-table will update Q-values for each power allocation under each channel condition based on its corresponding throughput. A certain power allocation will obtain a high Q-value if it achieves high downlink throughput. Each time a power allocation is selected, its Q-value will be updated based on its past throughput, current throughput, and predicted future throughput. After several epochs, the HAP will mostly select the power allocation with the highest Q-value to pursue high downlink throughput. Consequently, with the convergence of the learning process, for each given channel state, the best power allocation will achieve the highest Q-value. Thus, for each downlink transmission, the HAP will learn the certain transmission power p_i for each user to employ downlink NOMA transmission. At the *uplink* layer, each IoT device is an agent that independently learns its uplink MDP action $\check{a}_i = [\delta_i, \rho_i]$ which includes uplink transmission probability and slot selection. That means, in each frame, each device will learn to select a certain uplink transmission slot and probability to transmit. Similar to the downlink layer, the Q-table will update the Q-value for each action-state pair until converges. Therefore, for a given channel state, the uplink transmission slot and transmission probability with the highest transmission rate will obtain the highest Q-value. In the *stateless* layer, the system determines the uplink frame size and downlink power split ratio. During the warm-up period, the system randomly determines the uplink frame size and the

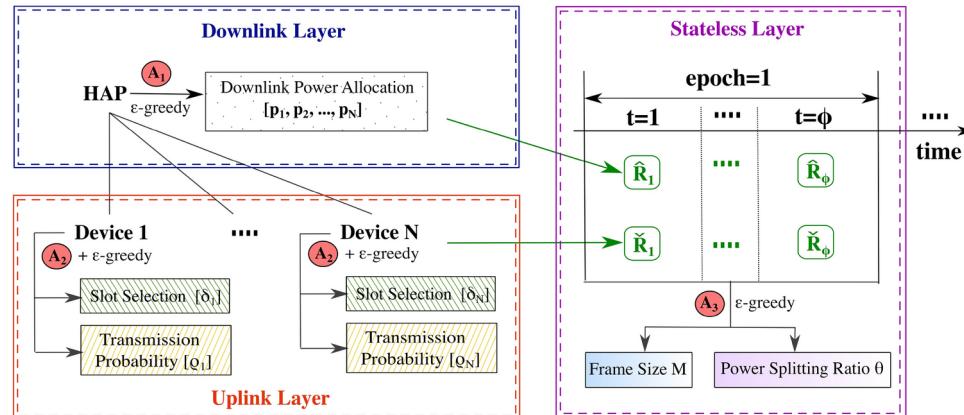


Fig. 3. Multi-Q includes downlink, uplink, and stateless layer. In the downlink layer, the HAP employs Algorithm 1, which is denoted as A_1 in the figure, to learn downlink power allocation. In the uplink, each user independently employs Algorithm 2, which is denoted as A_2 , to learn its own slot selection and transmission probability. Then the stateless layer collects the reward of both uplink and downlink for one epoch and then employs Algorithm 3 to determine the frame size and power split ratio.

downlink power-splitting ratio. For each frame size and power splitting ratio, the Q-table will update the Q-value based on the sum rate. After several epochs, the system will select the frame size and power-splitting ratio with the highest Q-value. The frame size and power splitting ratio that obtains a high system sum rate will get a high Q-value. Each time an action is selected, its Q-value will be updated based on its past reward, current reward, and possible future reward. Until convergence, the best frame size and power-splitting ratio will have the highest Q-value. Let Φ denote the number of epochs and ϕ denote the number of frames inside each epoch. Next, we present how each layer works.

Downlink layer

In the *downlink layer*, the HAP performs conventional Q-learning³³. Algorithm 1 demonstrates the steps of this layer. Each learning phase consists of ϕ time frames. Firstly, the HAP initializes its Q-table and learning parameter α and γ . During each time frame t , the HAP collects the channel condition h_i^t of user U_i to obtain its downlink state S_t , see line 5. After that, the HAP uses ϵ -greedy to select an action A_t which governs the transmission power allocation for all users. Concretely, with probability $(1 - \epsilon)$, the HAP selects the action A_t with the highest Q-value for state S_t , see line 10. After taking action A_t , each user collects its individual reward r_i^t and reports to the HAP. The HAP sums all rewards together to obtain the downlink reward R_t , see line 13. Then, the HAP observes its next state and finds the highest Q-value for the next state to update its Q-table.

```

1 Initialize learning parameter  $\alpha, \gamma$ 
2 Initialize each  $Q(S, A)$  randomly
3 for  $t \in \phi$  do
4   HAP collects  $h_i^t$  of all users
5   Obtain downlink state:  $S_t = \{h_1^t, h_2^t, \dots, h_N^t\}$ 
6   Generate a random number  $x$ 
7   if  $x < \epsilon$  then
8     | Select an action randomly
9   else
10    | Select an action  $A_t$  by solving:  $A_t(S_t) = \arg \max_{A \in \mathcal{A}^t} Q(S_t, A)$ 
11   end
12   Collect reward  $r_i^t$  of each user
13   Calculate downlink reward:  $R_t = \sum_{i=1}^M r_i^t$ 
14   Obtain the next downlink state:  $S_{t+1} = \{h_1^{t+1}, h_2^{t+1}, \dots, h_N^{t+1}\}$ 
15   Find  $\max_{A \in \mathcal{A}^{t+1}} \check{Q}(S_{t+1}, A)$ 
16   Update Q-value  $Q(S_t, A_t)$  as per Eq. (8)
17 end

```

Algorithm 1. Pseudocode for downlink Q-learning.

Uplink layer

In the uplink layer, each user acts as an agent to independently perform conventional Q-learning, see Algorithm 2. In each time frame $t \in \phi$, each user uses its channel condition h_i^t and battery level E_i^t as its current state $s_i^t = [h_i^t, E_i^t]$, see line 5. Based on ϵ -greedy, each user selects an action a_i^t that governs its transmission slot selection and transmission probability. Specifically, each user either selects an action randomly, see line 8, or selects the action with the highest Q-value, see 10. After that, each user observes its next state s_i^{t+1} , see line 13, and finds the corresponding highest Q-value, see 14. Then, each user updates its Q-table, and repeats the aforementioned steps.

```

1 Initialize learning parameter  $\alpha, \gamma$ 
2 Initialize each  $Q(s, a)$  to arbitrary values.
3 for  $t \in \phi$  do
4   for user  $i \in N$  do
5     Obtain current state  $s_i^t$ 
6     Generate a random number  $x$ 
7     if  $x < \epsilon$  then
8       | Select an action randomly
9     else
10      | Select an action  $a_i^t$  by solving:  $a_i^t(s_i^t) = \arg \max_{a \in \mathcal{A}^t} Q(s_i^t, a)$ 
11    end
12    Collect reward  $r_i^t$ 
13    Observe next state  $s_i^{t+1}$ 
14    Find  $\max_{a \in \mathcal{A}^{t+1}} \check{Q}(s_i^{t+1}, a)$ 
15    Update the Q-value  $Q(s_i^t, a_i^t)$  as per Eq. (8)
16  end
17 end

```

Algorithm 2. Pseudocode for Uplink Q-learning.*Stateless layer*

In the *stateless* layer, the learning phase consists of Φ epochs, see Algorithm 3. At the beginning of each epoch κ , the system selects an action $a_\kappa = [M, \theta]$ that governs the uplink frame size and downlink power split ratio. With probability ϵ , the system randomly selects an action, see line 5. Otherwise, the system will select an action with the highest probability, see line 9. After that, the system collects the reward for uplink and downlink, see line 11, 12, and accumulates downlink and uplink reward during epoch κ to obtain stateless reward, see line 13. Then, the system updates its Q-table and PMF. It then repeats the said steps.

```

1 Initialize learning parameter  $\lambda$ 
2 Initialize each  $Q(a)$  randomly
3 for  $\kappa \in \Phi$  do
4   Generate a random number  $x$ 
5   if  $x < \epsilon$  then
6     | Select an action randomly
7   else
8     | Select an action  $a_\kappa$  by solving:  $a_\kappa = \arg \max \Pr(a)$ 
9   end
10  Collect downlink reward during  $\kappa$  epoch  $R_\kappa^\downarrow$ 
11  Collect uplink reward during  $\kappa$  epoch  $R_\kappa^\uparrow$ 
12  Calculate joint reward:  $R_\kappa = R_\kappa^\downarrow + R_\kappa^\uparrow$ 
13  Update Q-value  $Q(a)$  as per Eq. (9)
14  Update PMF as per Eq. (10)
15 end

```

Algorithm 3. Pseudocode for Stateless Q-learning.**Evaluation**

We conducted our simulation using Matlab running on a machine with 8-Core Intel Core i9 @2.3 GHz with 16 GB of RAM. The path loss at reference distance 1 m is -20 dB²⁷. We fixed both the uplink length τ_{up} and downlink length τ_{down} to 1 s. We consider a packet size of $L = 21$ bytes as per the IPv4 standard, which includes 20 bytes for header and one byte of data. The average energy consumption rate ζ is 18 nJ/bit³⁵. Thus, the minimum energy consumption for transmission is $\varepsilon_0 = \zeta \times L = 3.024$ μ J. The battery capacity \mathcal{B} of each user is set to $5\varepsilon_0$. According to the non-linear model in³¹, we set the energy conversion efficiency parameters as $M = 0.02$ W, $a = 1500$, and $b = 0.0014$.

We compare Multi-Q against round-robin, TDMA, and Aloha. The round-robin protocol is used for both uplink and downlink transmissions; i.e., the HAP transmits downlink signals to each device in turn and devices

transmit to the HAP in turn. As for TDMA and Aloha, these protocols are for uplink transmissions only. Both these two protocols consider downlink NOMA with uniform power allocation. Then during uplink, TDMA assigns a dedicated time slot to each device. As for Aloha, devices with sufficient energy contend for an uplink time slot randomly. We measure and compare the performance of these protocols from three aspects including average system sum rate, average downlink transmission rate, and average uplink transmission rate. Apart from that, we study different HAP transmit power, device location, and power split ratio. Each simulation has 30,000 time frames. We collect the result in the last 300 frames after convergence, and plot the average of ten simulation runs. In terms of computational complexity, Multi-Q involves three layers of Q-learning. We analyze the computational complexity of each server at time t . For the downlink layer and uplink layer, each server needs to determine the Q-value of a state-action pair. This Q-value is calculated as per (8), which takes $O(1)$ time. For the stateless layer, a server needs to determine the Q-value of an action which is calculated as per (9). For each layer, the Q-value is updated only according to the reward of servers, which is calculated as per (5). Observe that (8), (9), and (5) only involve multiplication and addition operations. Moreover, Multi-Q is able to suit larger-scale networks. However, the downlink layer computational complexity may increase with a larger scale since it calls for global information on the server.

Convergence

To study convergence, users are placed at a distance of 1, 5, and 9 m from the HAP. We run our simulator for 200 iterations and each iteration contains 150 frames. We plot both uplink and downlink rates in Fig. 4. There is a warm-up period of 15,000 frames. Referring to Fig. 4, we can see that both uplink and downlink rates converged after 140 iterations. Concretely, the downlink rate converged to around 34 b/s/Hz, and the uplink rate converged to around 13.66 b/s/Hz.

Learning parameter

We now study learning parameters. Specifically, we study the uplink and downlink layer learning rate including uplink learning rate α_u and downlink learning rate α_d , the frame size and power ratio layer learning rate λ , discounting factor γ_d , and warm-up period. We can see from Fig. 5 that each learning parameter combination converged to a different sum rate. With a short warm-up period, the system will experience more randomness during convergence. The system converges faster when the frame size and power ratio layer uses a learning rate of λ .

HAP transmission power

We vary the transmission power of the HAP from 1 to 5 W. User are located at a distance of 1, 5, and 9 m to the HAP. The frame size is three for Aloha and TDMA. The path loss exponent is $n = 2.7$ ³⁶.

Figure 6a demonstrates the average sum rate of both uplink and downlink. The sum rate of uplink and downlink increases with a higher HAP transmission power for all methods. Concretely, both uplink and downlink rate increase with a higher HAP transmission power as shown in Fig. 6b. The reason for the increase in uplink rate is because users harvest more energy with a higher HAP transmission power. Thus users are able to transmit with a higher transmit power. Similarly, a higher HAP transmit power leads to a higher downlink rate.

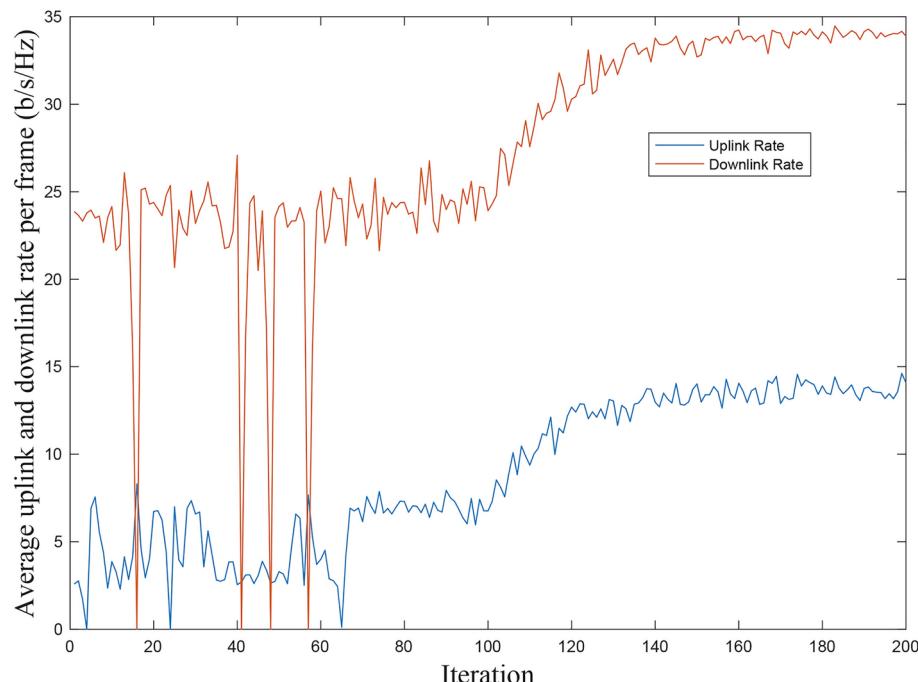


Fig. 4. Convergence curve for uplink and downlink.

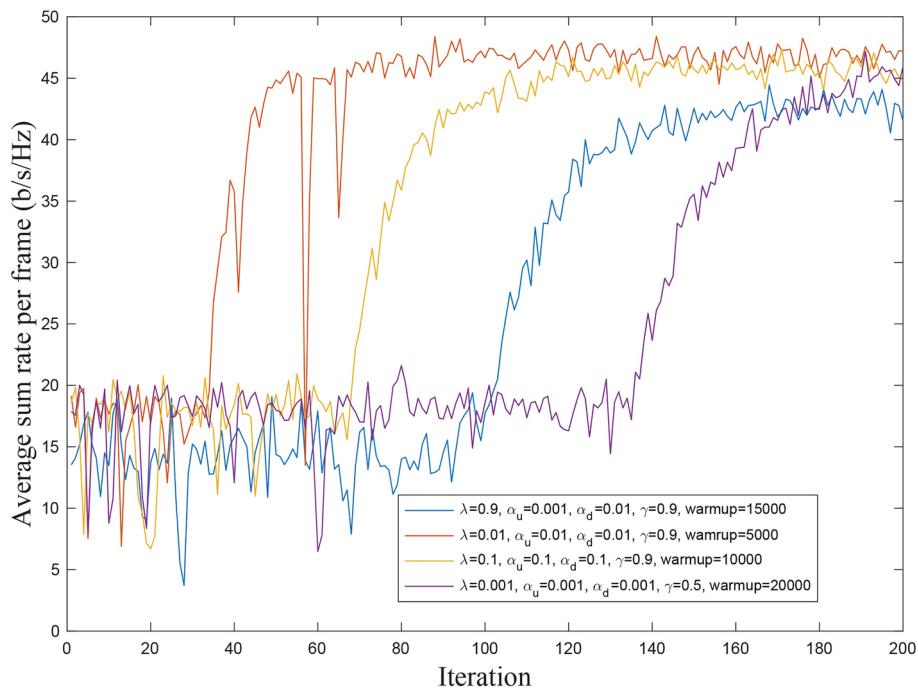


Fig. 5. Sum rate convergence curve for different learning parameters.

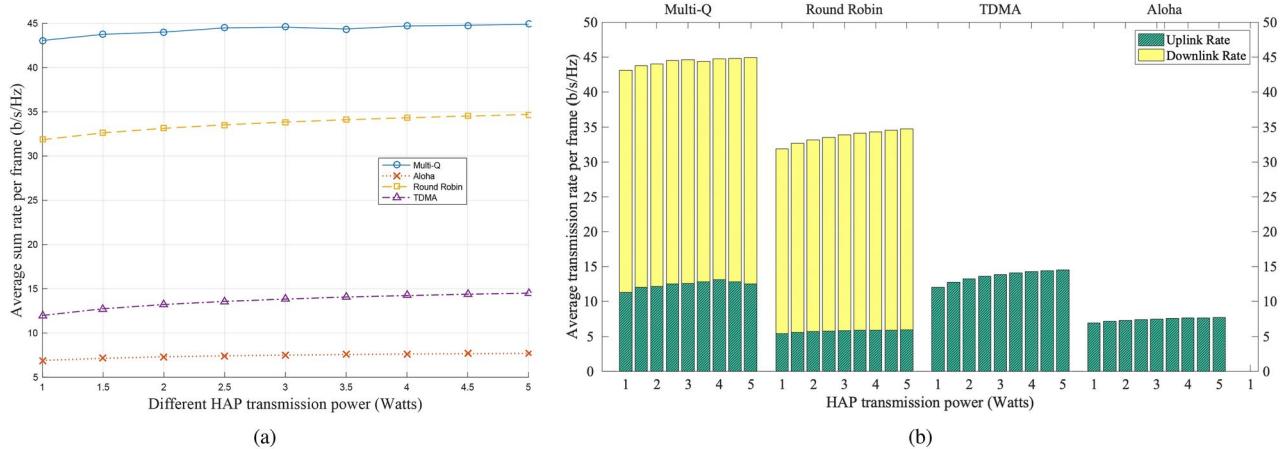


Fig. 6. The impact of different HAP transmission powers on the (a) average sum rate per frame and (b) average uplink rate and downlink rate per frame, respectively. The frame size for TDMA is three, and the path loss exponent is $n = 2.7$.

Multi-Q performs best when we vary the HAP transmission power from 1 to 5 W. From Fig. 6a, It is clearly that Multi-Q always reaches the highest sum rate from 1 to 5 W. Moreover, Multi-Q reaches an average of 44.3 b/s/Hz, which is 6× that of Aloha, 2.3× that of TDMA, and 30% more than round-robin. From Fig. 6b, Multi-Q reaches an average downlink transmission rate of 31.9 b/s/Hz which is the highest among all methods and achieves 6.6 b/s/Hz more than round-robin. TDMA and Aloha are even worse and just achieved zero downlink rates. The reason is because both TDMA and Aloha adopt uniform power distribution for each user during downlink. A uniform power distribution leads to decoding failures. Furthermore, Multi-Q performs better than round-robin for both uplink and downlink. Concretely, Multi-Q achieves an average of 12.4 b/s/Hz and 31.9 b/s/Hz rate for uplink and downlink, which is 6.6 b/s/Hz and 4.1 b/s/Hz more than that of round-robin, respectively. This is because Multi-Q users utilize the whole downlink period to receive data. On the other hand, for round-robin, users only receive data when it is polled by the HAP. In terms of downlink, round-robin experiences idle slots when users do not harvest sufficient energy. However, Multi-Q is able to avoid idle slots by dynamically adjusting the frame size and transmission probability based on the battery level and channel condition of users. Overall, Multi-Q performs significant advantages in terms of sum rate which is 6 times of

Aloha method. Simultaneously, Multi-Q also shows great advantages in terms of downlink rate which is 26% more than Round Robin and 100% more than TDMA and Aloha.

User location

We set five groups of user locations and the total distance from each group of devices to HAP is 15 m. We start from the group where each user is placed 5 m from the HAP. As channel gain disparity improves SIC decoding³⁷, we move users to different locations to obtain different channel gain conditions. Specifically, we consider five groups of user locations. The distance (in meters) of each user to the HAP is as follows: [5, 5, 5], [4, 5, 6], [3, 5, 7], [2, 5, 8], and [1, 5, 9]. The frame size is three for Aloha and TDMA. The path loss exponent is $n = 2.7$ ³⁶.

As shown in Fig. 7a, the average sum rate of both uplink and downlink increases when users have a more significant distance difference. This is because when we place users at different distances to the HAP, users experience significant differences in channel gains. Thus, the energy harvested by users vary considerably. This also means there will be one user located close to the HAP who transmits at a high power while another user located farther from the HAP that uses a low transmit power. This difference in transmit power helps increase the number of SIC decoding successes.

Multi-Q outperforms all other methods, especially when the distance between users is large. Specifically, Multi-Q achieves an average sum rate of 32 b/s/Hz for different locations. Simultaneously, Aloha, Round Robin, and TDMA achieve an average sum rate of 2.5 b/s/Hz, 29 b/s/Hz, and 4 b/s/Hz, respectively. When users are located at a distance of 1 m, 5 m, and 9 m to the HAP, Multi-Q achieves a sum rate of 44 b/s/Hz, which is six times that of Aloha, three times higher than TDMA, and 30% more than round-robin. The reason why Multi-Q performs better is because both Aloha and TDMA obtain zero downlink rate as shown in Fig. 7b. As both Aloha and TDMA employ uniform power distribution in the downlink, they always experience failures during downlink transmissions. When compared to round-robin, Multi-Q outperforms round-robin because Multi-Q simultaneously learns the frame size and transmission probability that avoid idle slots. Overall, Multi-Q always achieves the highest sum rate for different user locations which is an average of 11 times Aloha, 7 times TDMA and 10% more than Round Robin. Besides, Multi-Q shows great advantages in terms of downlink rate which is 100% higher than TDMA and Aloha, and 5% higher than Round Robin.

Power split ratio

Users are located at a distance of 1, 5, and 9 m to the HAP. The HAP transmission power is 3 W and the path loss exponent is $n = 2.7$. We vary the split ratio from zero to one with a step size of 0.1. Initially, the power split ratio is zero, meaning all received power is for energy harvesting. After that, we increase the power split ratio to 0.1. Thus, there is 10% power redirected for data reception and the remaining 90% is used for energy harvesting. Then we increase the power split ratio in steps of 0.1 until it reaches 1.0. Referring to Fig. 8a, Multi-Q achieves a sum rate around 44 b/s/Hz; it is able to converge to the optimal power split ratio starting from any initial ratio. The achieved sum rate of round-robin continues to rise until the power split ratio increases to 0.9. This is because when we increase the power split ratio, there will be more power distributed for downlink data reception and less power for energy harvesting. From Fig. 8b round-robin obtains an average of 25.3 b/s/Hz rate for downlink and 5.7 b/s/Hz for uplink. The downlink rate of round-robin is approximately 5× its uplink rate. Therefore, even if the uplink rate of round-robin decreases, the sum rate increases since the downlink rate increases more than the uplink rate. As the power split ratio increases from 0.9 to 1.0, the sum rate of round-robin decreases since there is no power distributed for downlink rate. However, Aloha and TDMA experience a decrease when we increase the power split ratio from 0 to 1.0. The reason is because users employ uniform power allocation for downlinks. Thus each user fails to decode the received packet since there is no difference between the transmit for each

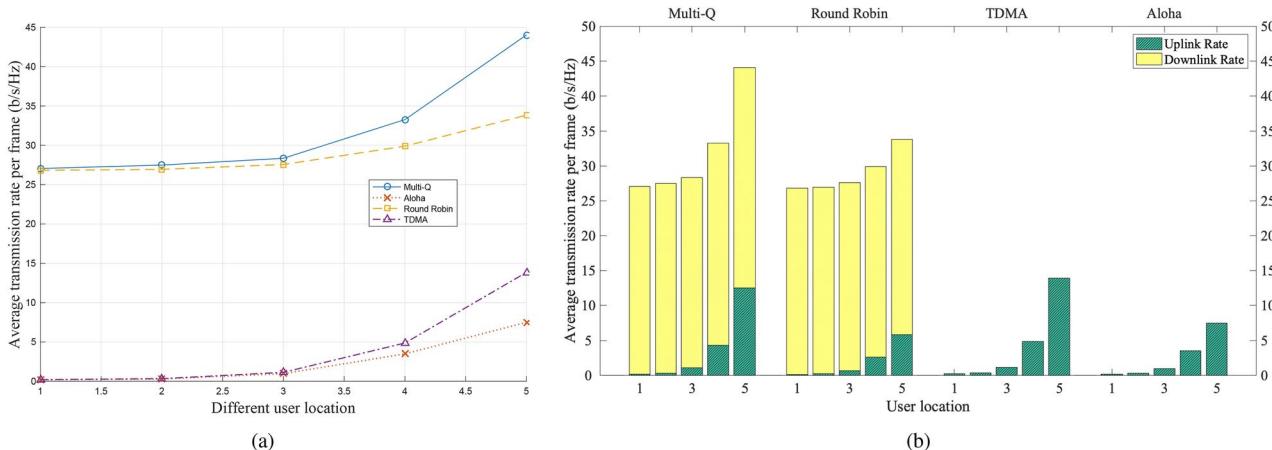


Fig. 7. The impact of different user locations on the (a) average sum rate per frame and (b) average uplink rate and downlink rate per frame, respectively. The HAP transmission power is 3 W. The frame size for TDMA and Aloha is three, and the path loss exponent is $n = 2.7$.

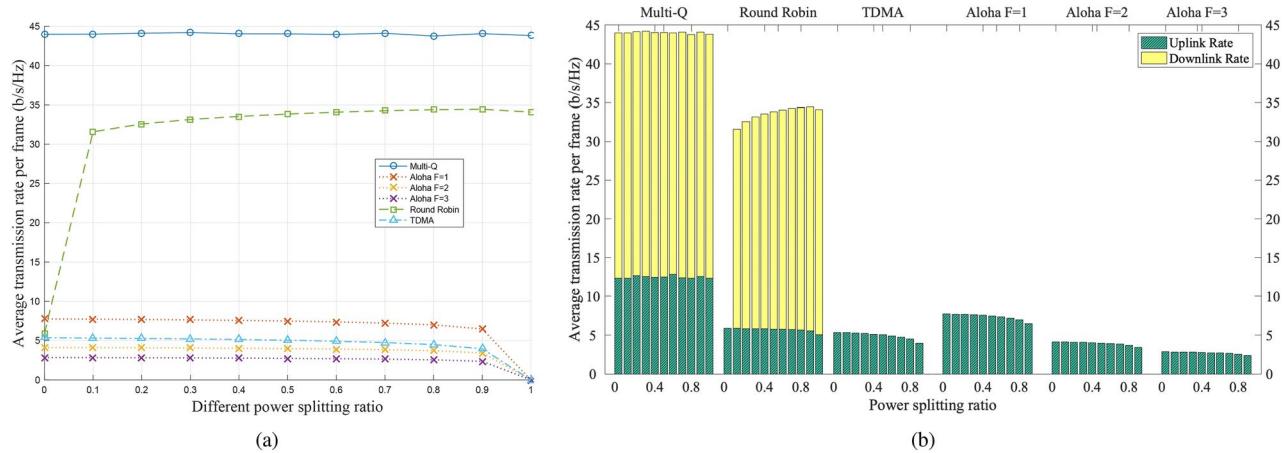


Fig. 8. The impact of different power split ratios on the (a) average sum rate per frame and (b) average uplink rate and downlink rate per frame, respectively. The HAP transmission power is 3 W. The frame size for TDMA and Aloha is three, and the path loss exponent is $n = 2.7$.

user. Further, for higher power split ratios, resulting in less harvested energy or transmit power, the uplink rate is appreciably lower.

We have also studied the performance of different frame sizes for Aloha. As shown in Fig. 8a, Aloha achieves its highest sum rate when the frame size is one. Although a larger frame size means fewer collisions, frame size one performs better than a larger frame size since we consider SIC, which allows decoding of multiple transmissions in the same slot. Moreover, a smaller frame size lengthens the transmission period, thus users are able to transmit longer.

Among all methods, Multi-Q performs best. Multi-Q achieves an average of 44 b/s/Hz sum rate, which is 9.8 times more than TDMA, 6.5 times that of Aloha when the frame size is one, and 1.4 times higher than round-robin. This is because Multi-Q simultaneously learns the power split ratio, frame size, uplink transmission probability, uplink slot selection, and downlink power allocation for all users. Specifically, Multi-Q learns the best frame size and slot selection to avoid uplink decoding failure and idle. Learning downlink power allocation for each user enhances decoding success. It also learns the power split ratio to balance uplink and downlink rates. Overall, Multi-Q shows great advantages in terms of sum rate over all other methods for different pre-set split ratios. Moreover, Multi-Q always achieves the highest downlink rate and uplink rate, which is 31.5 b/s/Hz and 12.5 b/s/Hz, respectively.

Conclusion

This paper has studied joint uplink and downlink transmissions in a wireless powered network that uses FSA and NOMA. In this respect, it has outlined a novel solution called Multi-Q that allows an HAP and devices to learn the optimal transmission policy. Specifically, for each system state, the HAP learns to optimize its transmit power allocation and frame size for uplinks. Similarly, devices learn the optimal power split ratio and transmission probability for each frame size. Advantageously, Multi-Q does not assume non-causal information and state transition probability. The simulation results show that Multi-Q achieves an average sum rate of 44 b/s/Hz which is 6 \times that of Aloha, 2.3 \times that of TDMA, and 30% more than round-robin. This is because our learning-based method Multi-Q can flexibly schedule the system to respond to different network conditions. Consequently, Multi-Q is able to obtain the best transmission strategy when compared to Aloha, TDMA, and so forth. Our work can be effortlessly extended to real-world IoT deployments like smart agriculture, smart transportation, smart cities, and so forth. This would better provide charging solutions and communications for agricultural sensors, parking sensors, and the like. As future work, we also aim to investigate whether our approach can be applied in multi-hop networks that include RF-energy harvesting relay nodes and extend our approach to be suitable for moving end devices such as moving vehicles. Moreover, an interesting future work is to study the performance of an approach based on deep Q-learning and/or an actor-critic.

Data availability

The datasets used and/or analysed during the current study available from the corresponding author on reasonable request.

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Author contributions

Y.L. conceptualized the study, conceived the experiment, and initial manuscript drafting; Y.L., Y.M., and W.W. contributed to data collection and analysis; All authors were involved in critically revising the manuscript, and approved the final manuscript.

Declarations

Competing interests

The authors declare no competing interests.

Additional information

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